

Single CDBA Based Voltage Mode Bistable Multivibrator and Its Applications

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Abstract

In this paper, current differencing buffered amplifier (CDBA) based bistable multivibrators are introduced. Each presented circuit is constructed using single CDBA as the basic active building block and three resistors. Two applications namely an astable and a monostable multivibrator are also realized to demonstrate the usefulness of the proposed bistable multivibrators. The presented circuits are simulated using PSPICE from Cadence Orcad16.2 to verify their functionality. Simulation results agree well with the theoretical analysis.

Keywords

Schmitt Trigger, Bistable Multivibrator, CDBA, Monostable Multivibrator

1. Introduction

Inherent wide bandwidth which is virtually independent of closed loop gain, greater linearity, and large dynamic range are the key performance features of current mode technique [1]. The CDBA is one such active element which inherits these advantages. In addition, it is free from parasitic capacitances [2] and hence is appropriate for high frequency operation. It provides further flexibility to the designers, enabling a variety of circuit designs, as it can operate in both current and voltage mode [3].

Bistable multivibrator, commonly known as Schmitt trigger, finds extensive applications in the fields of communication systems, instrumentation measurement systems, and power conversion control circuits [4]. It is commonly employed in monostable multivibrator [4]-[7], square wave generator [8]-[12], pulse width modulator (PWM) [13] [14], etc. Several implementations of the Schmitt triggers using different high-performance active building blocks have been proposed in open literature [15]-[19]. Conventional voltage-mode bistable multi-

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vibrators [15] employ an op-amp with a positive feedback. Current mode building blocks based voltage output bistable multivibrators are presented in [16]-[19]. Schmitt trigger based on two operational transconductance amplifiers (OTAs) and two resistors is presented in [16] wherein the output amplitude and threshold level can be independently/electronically tuned. Schmitt triggers based on current conveyors are presented in [17] [18] which use only single active element and their outputs are temperature-insensitive. Bistable multivibrator configurations using single operational transresistance amplifier (OTRA) are proposed in [19] which provide both Clockwise (CW) and counter clock wise (CCW) hysteresis functions. A comparative statement of the existing voltage mode schmitt triggers is reported in **Table 1**.

It may be observed from the table that

- the op-amp based structures [15] though provide voltage output at appropriate impedance level yet the constant gain-bandwidth product and lower slew rate of the op-amps limit their high frequency operations.
- the structure proposed in [16] provides temperature sensitive output
- the configurations of [16]-[18] provide voltage output at high impedance and hence require a buffer to drive the voltage input circuits. This increases the component count in the circuit.
- the structures presented in [17] [18] provide only CW hysteresis
- the OTRA based structures[19] can be used both for voltage and current inputs, however output can be only voltage type
- the CDBA based structure provides further flexibility as it can be driven by both voltage and current inputs and can provide both voltage and current outputs

Above discussion suggests that CDBA based design is one of the most suitable choice. To the best of authors' knowledge no CDBA based schmitt trigger circuit is available in literature. Thus this paper aims at introducing new CW and CCW Schmitt Triggers, using single CDBA and three resistors which will provide further flexibility to circuit designers. The PSPICE simulation results are also shown, which are in correspondence to the theoretical analysis. To show the usefulness of the presented circuits, the applications of the Schmitt triggers as square wave/triangular wave generator and monostable multivibrator are introduced.

The remaining paper is organized as follows. In Section 2 the function of a CDBA is introduced followed by the description of proposed circuits. The PSPICE simulations and experimental results to investigate the circuit performances are presented in Section 3 which are in confirmation with the theoretical propositions. In Section 4, application examples of the proposed circuits are given. The concluding remarks are presented in Section 5.

2. Circuit Description

The circuit symbol of CDBA is shown in **Figure 1** and the port characteristics are given by Equation (1)

$$\begin{bmatrix} I_z \\ V_w \\ V_p \\ V_n \end{bmatrix} = \begin{bmatrix} 0 & 0 & 1 & -1 \\ 1 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 \end{bmatrix} \begin{bmatrix} V_z \\ I_w \\ I_p \\ I_n \end{bmatrix} \tag{1}$$

Table 1. Comparison of existing voltage mode schmitt triggers.

Ref.	No. of active blocks used	No. of passive components	Hysteresis type	Output type	Output Impedance	Temperature sensitivity
[15]	1 Op-amp	3R	CW, CCW	Voltage	Low	No
[16]	2 OTA	2R	CW	Voltage	High	Yes
[17]	1 CC II+	3R	CW	Voltage	High	No
[18]	1CC II	3R	CW	Voltage	high	No
[19]	1 OTRA	2R	CW, CCW	Voltage	Low	No
Proposed	1 CDBA	3R	CW, CCW	Voltage	Low	No

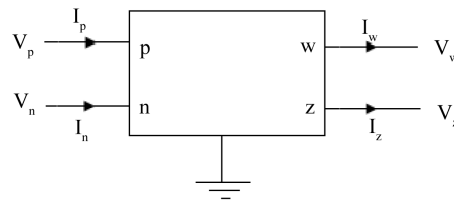


Figure 1. Block diagrammatic representation of CDBA.

2.1. The CW Schmitt Trigger

The proposed CW Schmitt Trigger configuration is shown in **Figure 2(a)**. The Input is provided through a small resistance R_1 at n terminal of the CDBA and output v_o is taken across w terminal. A high value resistance R_Z is connected at the z terminal of the CDBA which forces the circuit into saturation. Resistor R_2 forms a positive feedback loop to the ‘p’ input of the CDBA. Thus, the CDBA output saturates either at $+V_{sat}$, the positive saturation level or at the negative saturation level $-V_{sat}$. This circuit realizes the CW hysteresis characteristic as shown in **Figure 2(b)**.

For the CW hysteresis operation, the output V_o is initially assumed to be at the positive saturation level V_{sat}^+ . The current I_p and I_n of CDBA are given by

$$I_p = \frac{V_o}{R_2} \quad (2)$$

$$I_n = \frac{V_i}{R_1} \quad (3)$$

As V_i increases from zero, V_o remains at V_{sat}^+ until V_i reaches the upper threshold voltage V_{TH} thereby changing the output level from V_{sat}^+ to V_{sat}^- . This output level is maintained as long as V_i is greater than the lower threshold voltage V_{TL} . Assuming that V_o is at V_{sat}^+ and V_i is smaller than V_{TH} initially, I_p can be determined from Equation (3) as

$$I_p = \frac{V_{sat}^+}{R_2} \quad (4)$$

As v_i increases, current I_n gets closer to I_p and when I_n exceeds I_p , the output V_o switches to its negative saturation level V_{sat}^- . From Equations (3) and (4), the upper threshold voltage V_{TH} can be computed when I_p is equal to I_n and can be expressed as

$$V_{TH} = V_{sat}^+ \frac{R_1}{R_2} \quad (5)$$

The current of p terminal can now be computed as

$$I_p = \frac{V_{sat}^-}{R_2} \quad (6)$$

However, I_n remains same as Equations (3). By equating I_p and I_n the lower threshold voltage V_{TL} can be determined as

$$V_{TL} = V_{sat}^- \frac{R_1}{R_2} \quad (7)$$

The output level will switch back to V_{sat}^+ once I_p gets more positive than I_n .

2.2. The CCW Schmitt Trigger

The CCW Schmitt Trigger is shown in **Figure 3(a)** wherein, the input voltage is connected at p terminal of CDBA. The currents I_p and I_n can be computed as

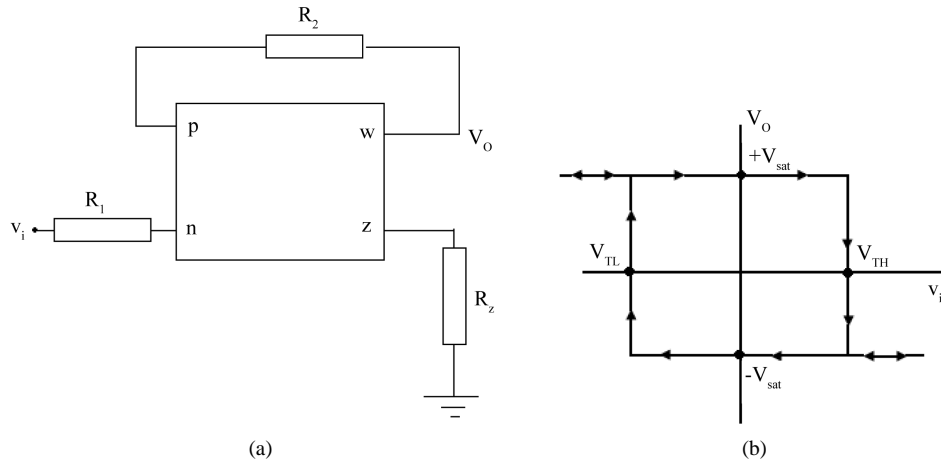


Figure 2. (a) CW Schmitt Trigger; (b) CW Hysteresis Curve.

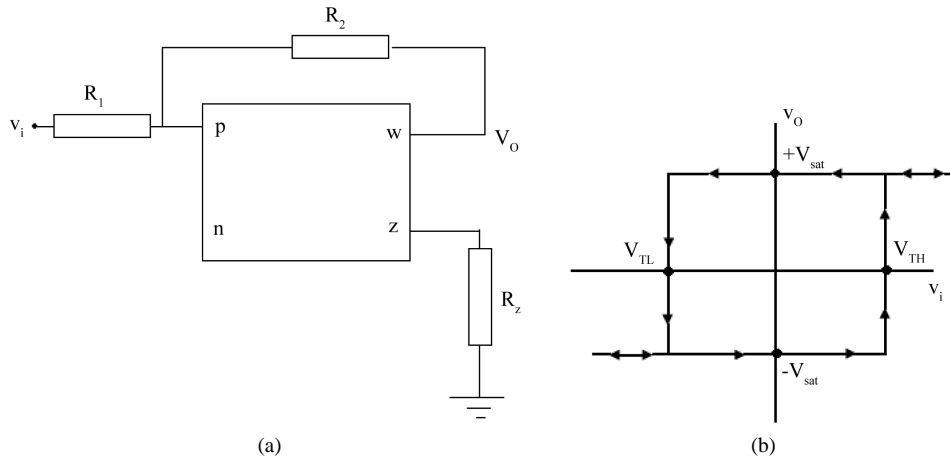


Figure 3. (a) CCW Schmitt Trigger; (b) CCW Hysteresis Curve.

$$I_p = \frac{V_i}{R_1} + \frac{V_o}{R_2} \quad (8)$$

$$I_n = 0 \quad (9)$$

Assuming that \$V_o\$ is at \$V_{sat}^+\$ and \$V_i\$ is initially larger than \$V_{TL}\$, then \$V_{TL}\$ can be computed as

$$V_{TL} = V_{sat}^+ \frac{R_1}{R_2} \quad (10)$$

When \$V_i\$ is smaller than \$V_{TL}\$, output \$V_o\$ switches to \$V_{sat}^-\$. With increasing \$V_i\$, \$I_p\$ also increases and forces the output to change its state when \$I_p\$ exceeds \$I_n\$. The upper threshold voltage \$V_{TH}\$ can thus be derived as

$$V_{TH} = V_{sat}^- \frac{R_1}{R_2} \quad (11)$$

Hysteresis Curve for CCW Schmitt Trigger is shown in Figure 3(b).

2.3. Schmitt Trigger with Reference Voltage

For the circuits shown in Figure 2(a) and Figure 3(a) the \$V_{TH} = -V_{TL}\$ and hence the switching voltage (\$V_{ST}\$) defined as \$(V_{TH} + V_{TL})/2\$ is zero. Some applications require that \$V_{TH}\$ and \$V_{TL}\$ both should either be positive or nega-

tive resulting in finite value of V_{ST} . This can be accomplished by adding a reference voltage to the circuit of **Figure 2(a)** and **Figure 3(a)** which results in following four configurations

1. CW Schmitt Trigger with positive V_{ST} , shown in **Figure 4(a)**,
2. CW Schmitt Trigger with negative V_{ST} , depicted in **Figure 4(b)**,
3. CCW Schmitt Trigger with positive V_{ST} , given in **Figure 5(a)**,
4. CCW Schmitt Trigger with negative V_{ST} , shown in **Figure 5(b)**.

For the circuit of **Figure 4(a)** I_p and I_n are given as

$$I_p = \frac{V_o}{R_2} + \frac{V_{dc}}{R_s} \tag{12}$$

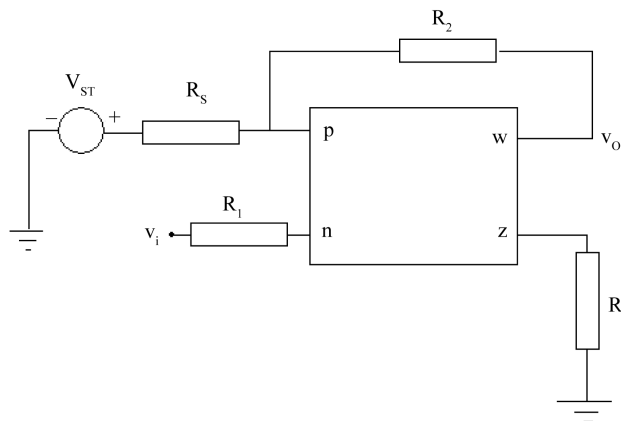
$$I_n = \frac{V_i}{R_1} \tag{13}$$

Using routine analysis the V_{TH} and V_{TL} can be computed as

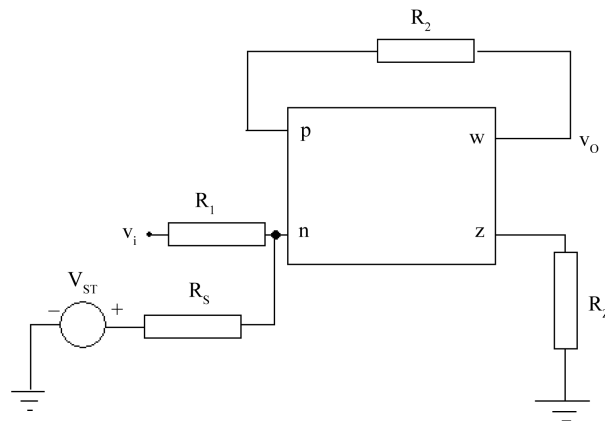
$$V_{TH} = V_{ST} + V_{sat}^+ \frac{R_1}{R_2} \tag{14}$$

$$V_{TL} = V_{ST} + V_{sat}^- \frac{R_1}{R_2} \tag{15}$$

where



(a)



(b)

Figure 4. (a) CW Schmitt Trigger with positive V_{ST} ; (b) CW Schmitt Trigger with negative V_{ST} .

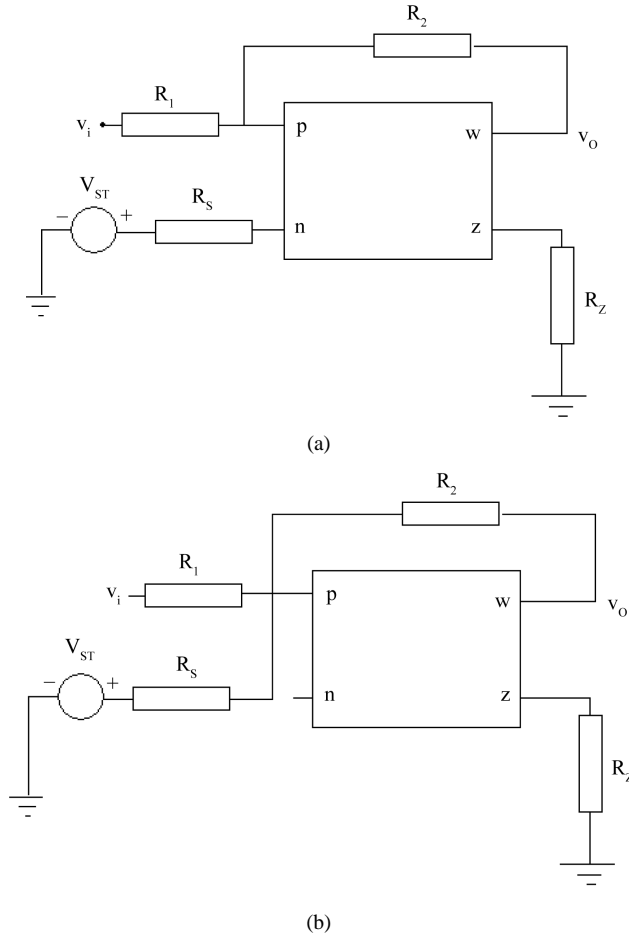


Figure 5. (a) CCW Schmitt Trigger with positive V_{ST} ; (b) CCW Schmitt Trigger with negative V_{ST} .

$$V_{ST} = \frac{R_1}{R_S} V_{dc} \quad (16)$$

Similarly the V_{TH} and V_{TL} for negative switching, as given in **Figure 4(b)**, can be derived as

$$V_{TH} = -V_{ST} + V_{sat}^+ \frac{R_1}{R_2} \quad (17)$$

$$V_{TL} = -V_{ST} + V_{sat}^- \frac{R_1}{R_2} \quad (18)$$

For CCW Schmitt Trigger with positive switching voltage, the threshold voltages V_{TH} and V_{TL} are given by Equations (14) and (15) respectively whereas for CCW configuration with negative switching voltage are given in Equations (17) and (18) respectively.

3. Simulation and Experimental Results

To validate the theoretical predictions, the proposed bistable multivibrator circuits have been simulated using PSPICE. The CDBA is realized using current feedback operational amplifier (CFOA) IC AD 844 as shown in **Figure 6** [2]. PSPICE Macro model of CFOA IC AD 844AN [20] is used for simulations and supply voltages used are ± 10 V.

Figure 7(a) shows the simulation results of CW Configuration for $R_Z = 500$ k Ω , $R_1 = 5$ k Ω , $R_2 = 10$ k Ω . The

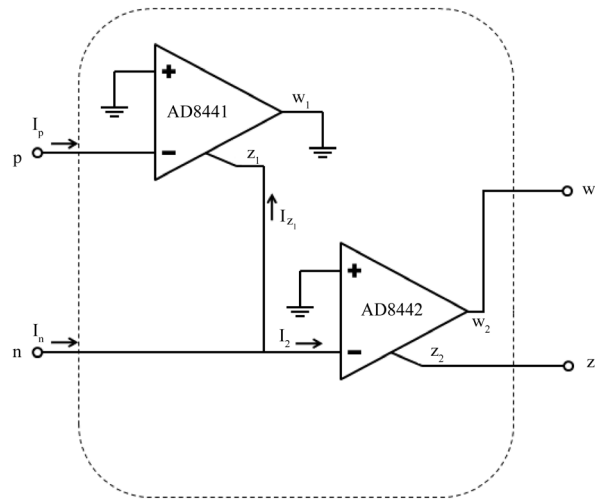
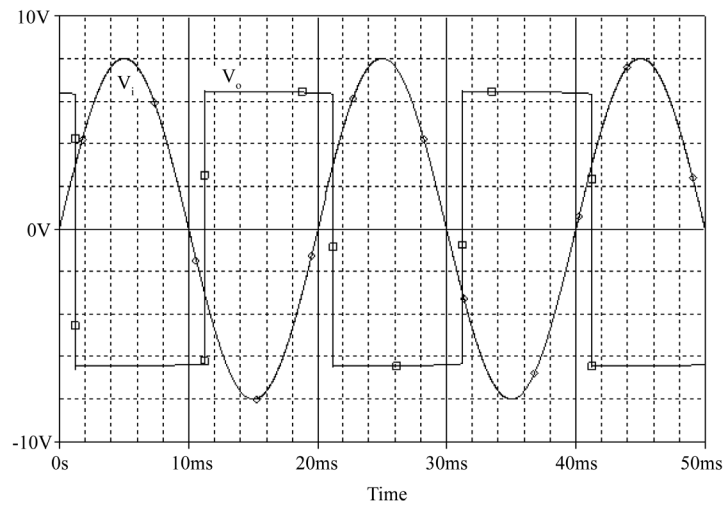
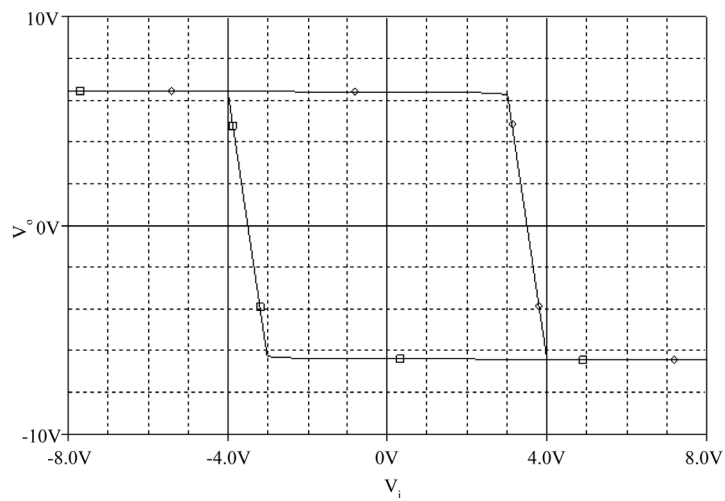


Figure 6. CFOA based implementation of CDBA [2].



(a)



(b)

Figure 7. (a) Output of the CW Schmitt Trigger; (b) CW Hysteresis Curve.

saturation levels are ± 6.3 V. The input voltage is a 50 Hz sinusoid with signal swing from -8 V to $+8$ V. The simulated threshold levels are ± 3 V which are in accordance with the theoretically computed value of ± 3.15 V. **Figure 7(b)** shows the hysteresis curve for CW configuration. The results for CCW configuration are depicted in **Figure 8** wherein, the component values and supply voltages are chosen same as that for the CW operation.

The transient responses for CW and CCW configurations are shown in **Figure 9(a)** and **Figure 9(b)** respectively for an input frequency of 250 KHz.

The frequency response of CW Schmitt Trigger is shown in **Figure 10** having 3 dB bandwidth as 2.6 MHz.

Transient response for CCW for positive switching voltage is shown in **Figure 11(a)** for $V_{TH} = 6$ V and $V_{TL} = 2$ V and corresponding transfer characteristics is shown in **Figure 11(b)**. The component values are computed as $R_Z = 500$ k Ω , $R_1 = 3.17$ k Ω , $R_S = 4.6$ k Ω , $R_2 = 10$ k Ω and $V_{dc} = 5$ V. Transient response and hysteresis curve for CW configuration with negative switching voltage for $V_{TH} = 6$ V and $V_{TL} = 2$ V are shown in **Figure 12(a)** and **Figure 12(b)** respectively. The simulated results are in close agreement with theoretical values for both the configurations.

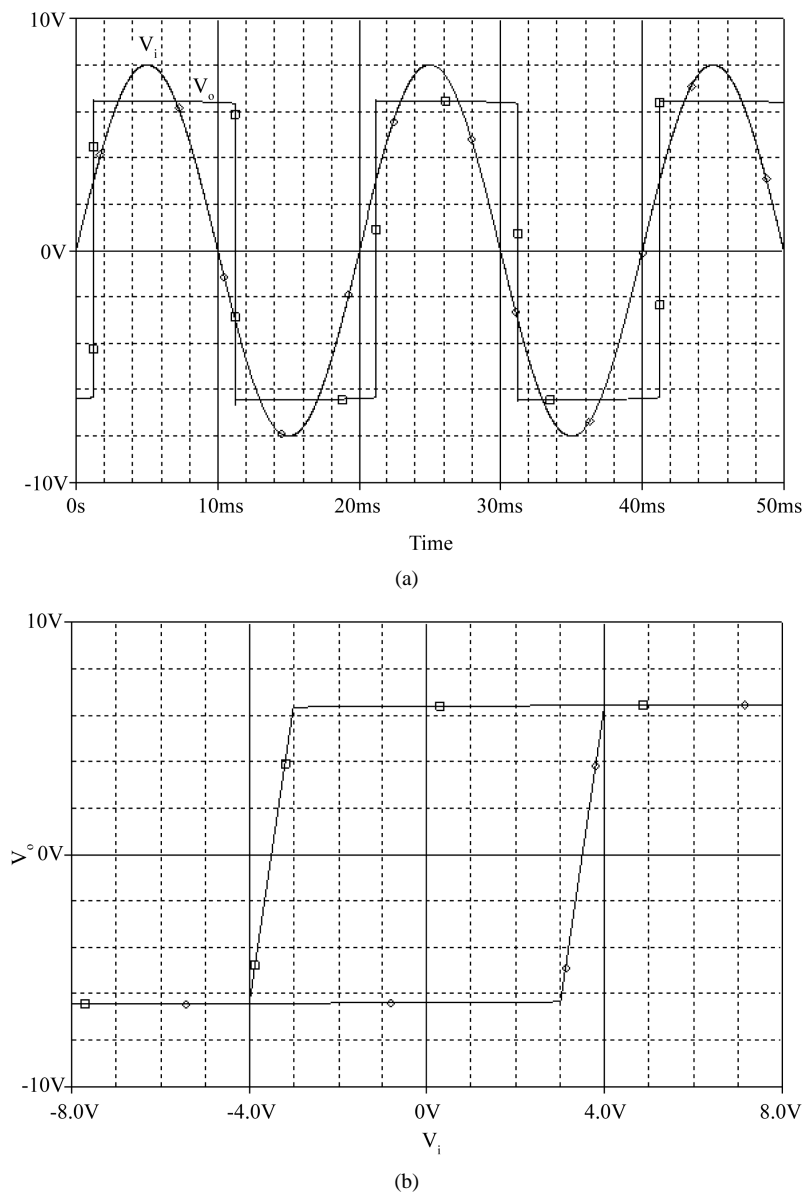
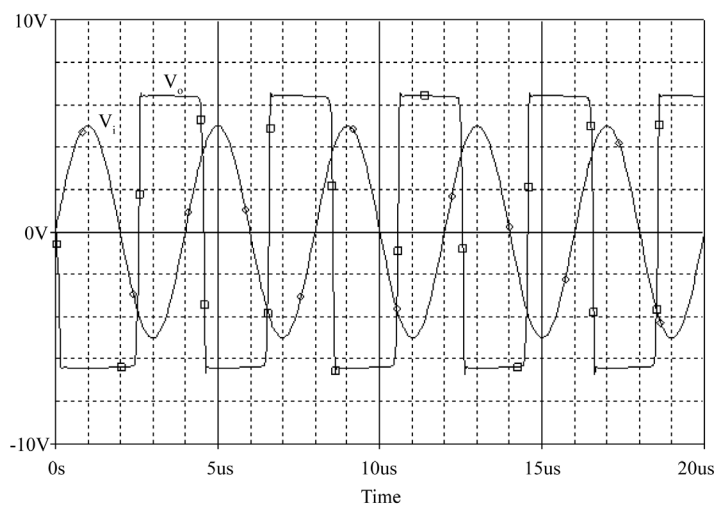
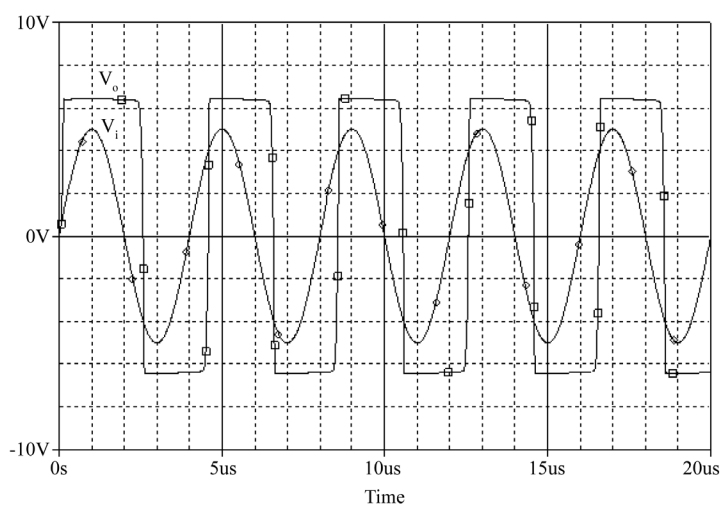


Figure 8. (a) Output of the CCW Schmitt Trigger; (b) CCW Hysteresis Curve.



(a)



(b)

Figure 9. Outputs for 250 KHz input signal (a) CW output; (b) CCW output.

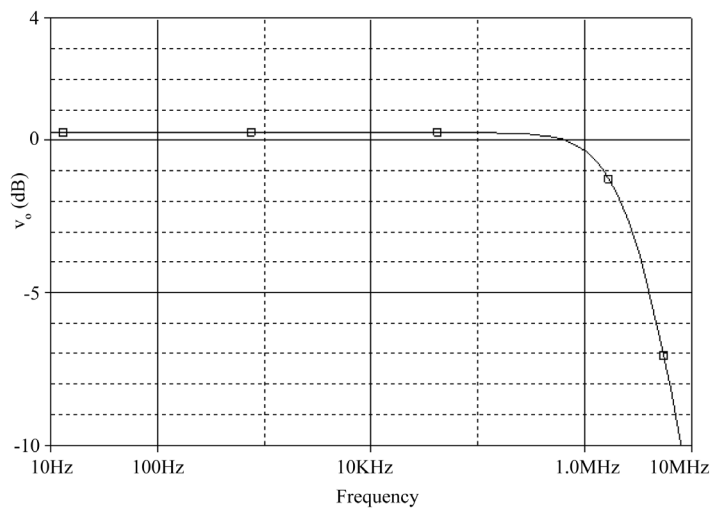


Figure 10. Frequency response of CW Schmitt Trigger.

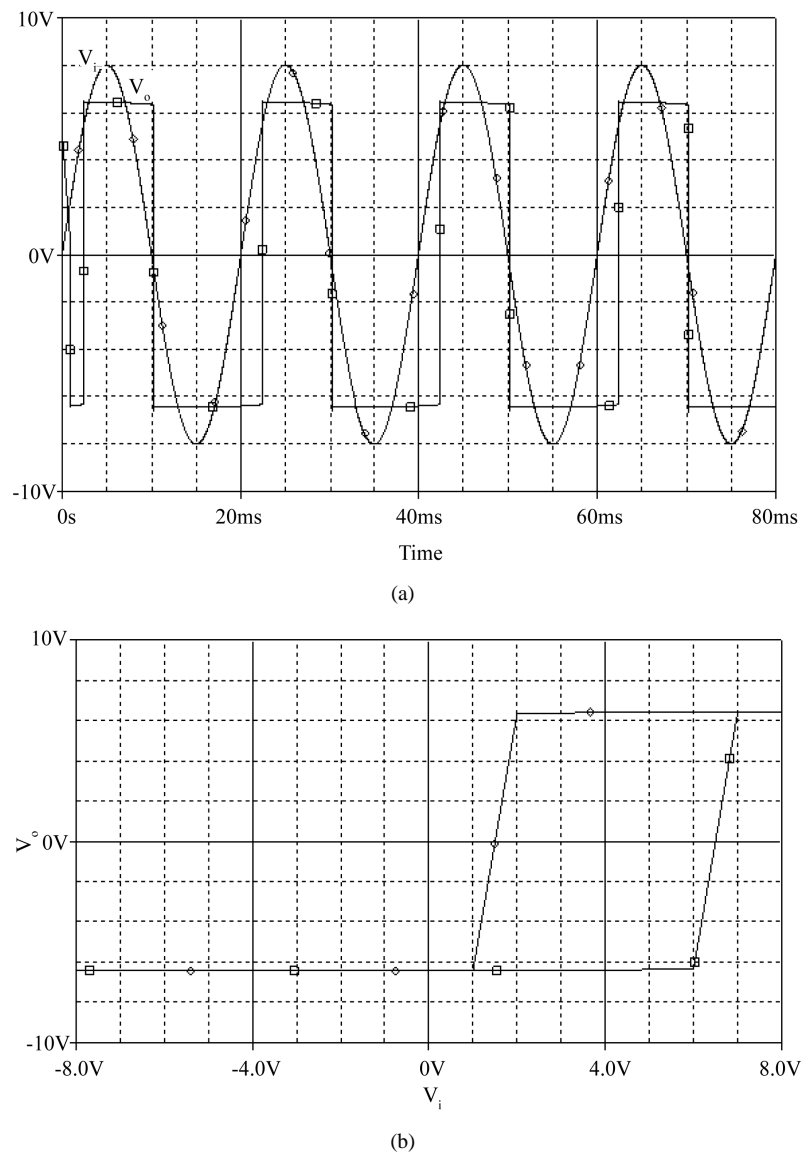
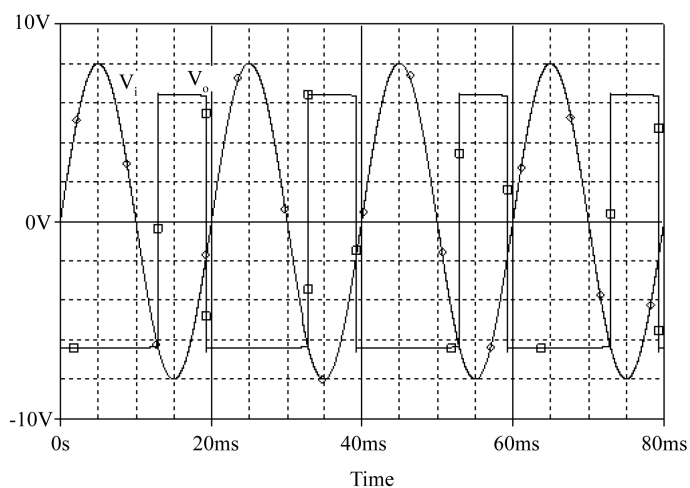


Figure 11. (a) CCW output with positive switching voltage; (b) CCW Hysteresis Curve with positive switching voltage.

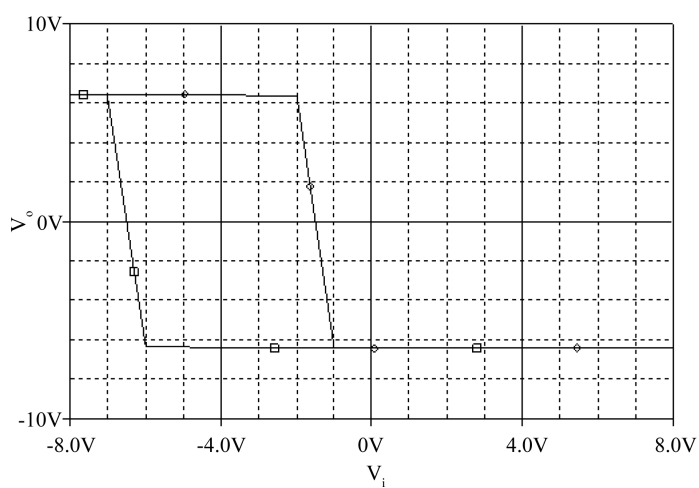
Figure 13(a) shows the experimental results of CW Configuration for $R_Z = 100\text{ k}\Omega$, $R_1 = 4.7\text{ k}\Omega$, $R_2 = 4.7\text{ k}\Omega$ with supply voltage of $\pm 12\text{ V}$. The saturation levels are $\pm 10\text{ V}$. The input voltage is a 1 KHz sinusoid with signal swing from -10 V to $+10\text{ V}$. The observed threshold levels are $\pm 10\text{ V}$ which are same as the theoretically computed value. **Figure 13(b)** shows the hysteresis curve for CW configuration and the experimental output for CCW configuration for an applied input voltage of 6 KHz sinusoid with signal swing from -10 V to $+10\text{ V}$ is presented in **Figure 13(c)**. The component values are chosen as $R_Z = 100\text{ k}\Omega$, $R_1 = 4.7\text{ k}\Omega$, $R_2 = 4.7\text{ k}\Omega$ and supply voltages are $\pm 8\text{ V}$. Observed saturation voltages are $\pm 6\text{ V}$ giving threshold voltages as $\pm 6\text{ V}$ and are equal to theoretical values.

4. Applications

In the following subsections two well known applications of Schmitt trigger namely triangular/square wave Generator and monostable multivibrators are developed to demonstrate the utility of proposed work in circuit applications.

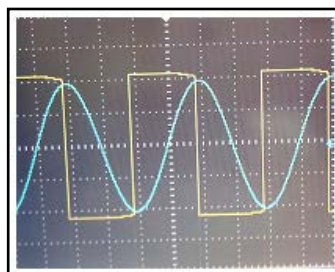


(a)

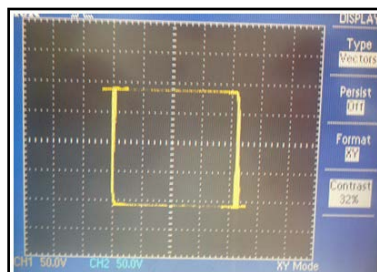


(b)

Figure 12. (a) CW output with negative switching voltage; (b) CW hysteresis curve with negative switching voltage.



(a)



(b)



(c)

Figure 13. Experimental results (a) output of CW Schmitt Trigger; (b) hysteresis curve for CW configuration; (c) output of CCW Schmitt Trigger.

4.1. Triangular/Square Wave Generator

The circuit of CDBA Schmitt trigger based triangular/square wave generator is shown in **Figure 14**. The circuit can be viewed as two cascaded blocks. The circuitry comprising of CDBA I is a Schmitt trigger, while the cir-

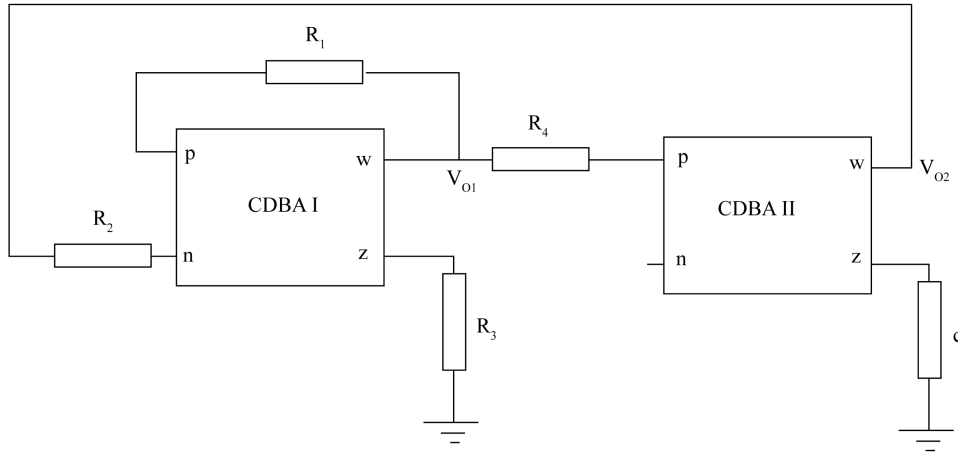


Figure 14. Triangular/square wave generator.

circuit comprising of CDBA II, is a simple integrator. The Schmitt trigger continuously compares the current I_{p1} and I_{n1} and accordingly the output V_{o1} swings repetitively between positive saturation level V_{sat}^+ and negative saturation level V_{sat}^- . Assuming initially the output V_{o1} to be at V_{sat}^+ , which is input to integrator, will charge the capacitor and would result in output voltage V_{o2} that is linearly rising. As a result the current I_{n1} will rise and when exceeds I_{p1} the output V_{o1} switches to V_{sat}^- . Now capacitor would begin to charge in opposite direction resulting in a negative ramp output at V_{o2} . As soon as I_{n1} falls below I_{p1} , V_{o1} switches back to V_{sat}^+ and V_{o2} become a positive going ramp again. For Schmitt trigger the V_{TH} and V_{TL} can be computed as $V_{sat}^+ R_2/R_1$ and $V_{sat}^- R_2/R_1$ respectively. Using the routine analysis the time period of the waveform can be computed as

$$T = 4V_{TH} R_4 C / V_{sat}^+ = 4R_2 R_4 C / R_1 \quad (19)$$

This gives frequency of oscillation as

$$f_o = 1/T = R_1 / 4R_2 R_4 C \quad (20)$$

The simulated square wave output V_{o1} and triangular output V_{o2} are shown in **Figure 15** for $R_1 = 10 \text{ k}\Omega$, $R_2 = 100 \text{ }\Omega$, $R_3 = 20 \text{ k}\Omega$, $R_4 = 5 \text{ k}\Omega$, and $C = 1 \text{ }\mu\text{F}$.

4.2. Monostable Multivibrator

The realization of the monostable multivibrator is shown in **Figure 16**. The positive feedback loop is completed using the capacitor and a resistor. Under stable state the V_c is clamped by diode. To ensure the stable-state operation, R_Z must be high enough to make output voltage V_o switch to positive saturation level V_{sat}^+ . Under stable state, ignoring the diode drop, the currents I_z and I_p are given by

$$I_z = V_{sat}^+ / R_Z \quad (21)$$

$$I_p = V_o / R_F = V_{sat}^+ / R_F \quad (22)$$

Now if a positive-edge triggering signal I_{trig} is applied at terminal n of CDBA, the circuits enter into the quasi-stable State. As I_n is more positive than I_p the output voltage V_o jumps to V_{sat}^- and C starts to discharge through R_F . In the quasi-stable state, the expressions of I_p and I_z are given by

$$I_z = V_z / R_Z = V_{sat}^- / R_Z \quad (23)$$

$$I_p = (V_{sat}^- - V_C) / R_Z \quad (24)$$

And capacitor discharging equation can be expressed as

$$v_C = V_{sat}^- \left(1 - e^{-\left(\frac{t-T_1}{R_F C}\right)} \right) \quad (25)$$

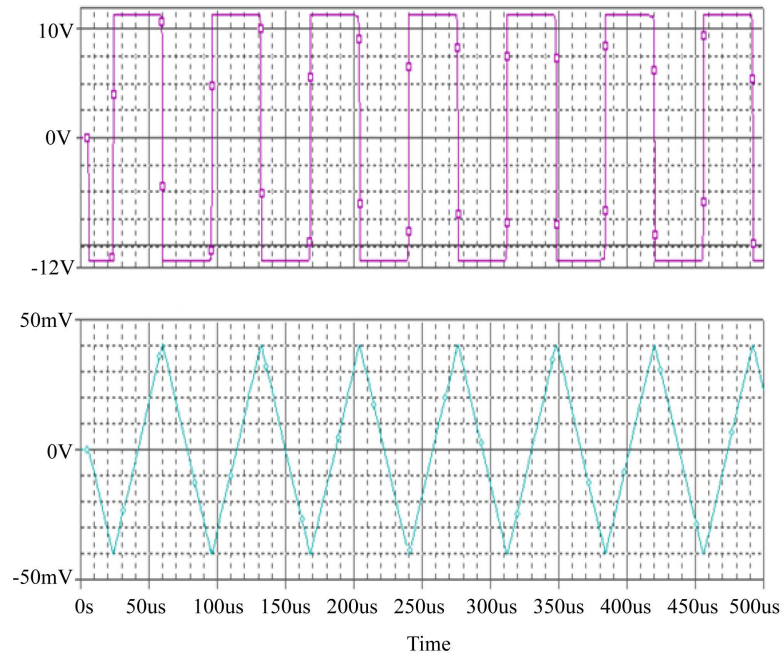


Figure 15. Output of triangular/square wave generator.

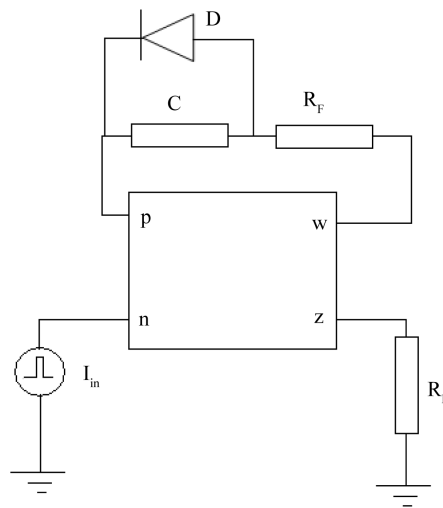


Figure 16. Monostable multivibrator.

at $t = T_2$ the capacitor voltage reaches the threshold voltage V_{TL} , when output voltage switches back to $V_{sat}^+ \cdot V_{TL}$ can be derived by equating Equations (23) and (24) and is given by

$$V_{TL} = V_{sat}^- (1 - R_F / R_Z) \tag{26}$$

From Equations (25) and (26) the pulse width $T (T_2 - T_1)$ for which the circuit remains in quasi stable sate can be computed as

$$T = R_F C \ln \frac{R_F}{R_Z} \tag{27}$$

Figure 17 shows the output of monostable multivibrator for $R_F = 20 \text{ k}\Omega$, $R_Z = 500 \text{ k}\Omega$, $C = 10 \text{ nF}$ having $T = 59 \text{ }\mu\text{s}$ as against calculated value of $64 \text{ }\mu\text{s}$.

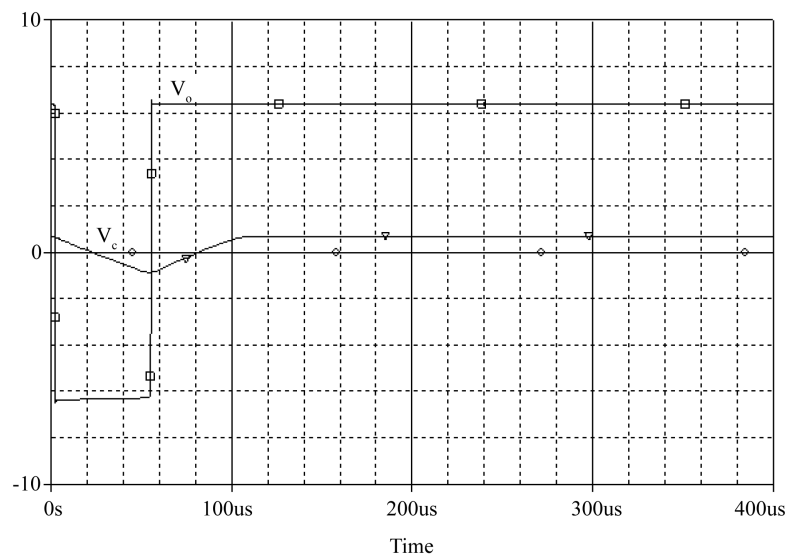


Figure 17. Output of monostable multivibrator.

5. Conclusion

In this paper single CDBA based bistable multivibrator configurations are proposed which include CW, CCW Schmitt Triggers with and without reference voltage. Two applications namely square wave/triangular wave generator and monostable multivibrator are realized to demonstrate the usefulness of the proposed bistable multivibrators. The simulation and experimental results are found to be in close agreement to theoretical predictions. The proposed configurations are one of the best choices for voltage mode applications. Also, due to inherent flexibility of signal usage in CDBA the proposed configurations can easily be extended to current/transimpedance/transadmittance mode depending upon the applications.

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